Lecture #2: LLSE — geometric interpretation

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1 Geometric Interpretation

Recall from Lecture #1, the llse. of x, \hat{x} is given by

$$\hat{x} = \Sigma_{xy} \Sigma_{yy}^{-1} y = K_o y$$

Interpretation:

$$K_o = \Sigma_{xy}\Sigma_{yy}^{-1} \Leftrightarrow K_o \, \mathsf{E}[yy^*] = \mathsf{E}[xy^*] \Leftrightarrow \mathsf{E}[(x - K_o y)y^*] = 0$$

Thus, if the random variables are viewed as vectors, with the inner product defined as $\langle x, y \rangle = \mathsf{E}[xy^*]$, then $\mathsf{E}[(x - K_o y)y^*]$ has the geometric interpretation

$$\langle x - K_o y, y \rangle = 0 \leftrightarrow x - K_o y \perp y,$$

that is, the estimation error is orthogonal to observations!

Two questions related to the geometric interpretation:

- 1. Which vector space?

 Hilbert space, i.e., a complete normed linear vector space.
- 2. Is $\langle x, y \rangle = \mathsf{E}[xy^*]$ a proper inner product?
 - Linearity: $\langle \alpha x + \beta y, z \rangle = \alpha \langle x, z \rangle + \beta \langle y, z \rangle$

$$\mathsf{E}[(\alpha x + \beta y)z^*] = \alpha \, \mathsf{E}[xz^*] + \beta \, \mathsf{E}[yz^*] \quad \textit{Check!}$$

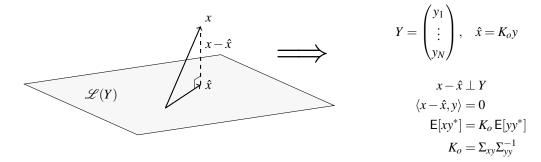
• Symmetry: $\langle x, y \rangle = \langle y, x \rangle^*$

$$E[xy^*] = E[(yx^*)^*] = E[yx^*]^*$$
 Check!

• Non-degeneracy: $\langle x, x \rangle \succeq 0$, *i.e.*, positive semi-definite, and $\langle x, x \rangle = 0$ if and only if x = 0. $E[xx^*] = \Sigma_{xx}$: Covariance matrices are per definition always positive (semi-)definite.

1.1 Geometric Derivation of LLSE

Find a vector \hat{x} in the linear space spanned by $\{y_1, \dots, y_N\}$ such that $||x - \hat{x}||$ is minimized.



2 Winer filtering (WF)

Given the observations $\{y_i\}_{i=-\infty}^k$, find the llse. of x_{k+n} .

Assumptions: x_k and y_k are scalar processes, that are jointly stationary with exponentially bounded (cross-)covarainace, *i.e.*, $|r_{xx}(k)| < K\alpha^{|k|} \forall k, K > 0, 0 < \alpha < 1$, then a spectrum exists.

That is, find

$$\hat{x}_{k+n} = \sum_{i=0}^{\infty} h_{k,i} y_{k-i},$$

where $h_{k,i}$ is possibly time varying filter coefficients subjects to $E[(x_{k+n} - \hat{x}_{k+n})^2]$ is minimized.

$$\underbrace{ \begin{array}{cccc} smoothing & filtering & prediction \\ \dots, y_{k-2}, y_{k-1}, & & & \\ \hline y_k & & & \\ \hline & & & k \\ n < 0 & & n = 0 & n > 0 \\ \end{array} }$$

Orthogonality properties gives, for all $j \le k$:

$$\begin{aligned} x_{k+n} - \hat{x}_{k+n} &\perp y_i, \\ \langle x_{k+n} - \sum_{i=0}^{\infty} h_{k,i} y_{k-i}, y_j \rangle &= 0, \\ \mathbb{E}[x_{k+n} y_j^*] &= \sum_{i=0}^{\infty} h_{k,i} \mathbb{E}[y_{k-i} y_j^*], \\ r_{xy}(k+n-j) &= \sum_{i=0}^{\infty} h_{k,i} r_{yy}(k-i-j). \end{aligned}$$

Make a change variable, $k - j \rightarrow \ell$, gives

$$r_{xy}(\ell+n) = \sum_{i=0}^{\infty} h_{\ell+j,i} r_{yy}(\ell-i), \quad \forall \ell \geq 0.$$

Note, neither $r_{xy}(\ell+n)$ nor $r_{yy}(\ell-i)$ depend on j; hence, $h_{\ell+j,i}=h_i$, i.e., the filter coefficients are time invariant.

Wiener-Hopf equation:

$$r_{xy}(\ell+n) = \sum_{i=a}^{b} h_i r_{yy}(\ell-i), \quad \forall \ell \geq 0.$$

If: • $\sum_{-\infty}^{\infty}$: Non-causual *Wiener filter* (WF, z-transform), easy to solve.

- \sum_{a}^{b} : FIR (finite impulse response) WF (linear system equation), easy to solve.
- \sum_{0}^{∞} : Casual WF.
- $\sum_{-\infty}^{-1}$: Anti-casual WF.

2.1 Spectral Factorization

The spectrum of a signal y_k

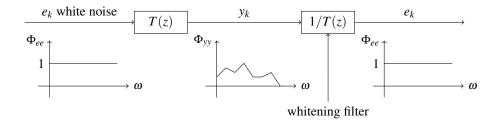
$$\begin{split} \Phi_{yy}(\omega) &= \tilde{\Phi}_{yy}(z) \Big|_{z=e^{i\omega}} \\ \tilde{\Phi}_{yy}(z) &= \mathscr{Z}\{\underbrace{r_{yy}(k)}_{\text{acf}}\} = \sum_{k=\infty}^{\infty} r_{yy}(k)z^{-k} \end{split}$$

Note, $r_{yy}(k) = r_{yy}(-k) \Rightarrow \tilde{\Phi}_{yy}(z) = \tilde{\Phi}_{yy}(z^{-1})$ which implies that $\tilde{\Phi}_{yy}(z)$ has symmetry with respect to mirroring in the unit circle. Hence, if $z = r_i$ has a pole (zero) in the unit circle, then $z = r_i^{-1}$ is also a pole (zero).

If $\tilde{\Phi}_{yy}(z)$ has no poles of zeroes on the unit circle, *i.e.*, $0 < \Phi_{yy}(\omega) < \infty \ \forall \omega$, then

$$\tilde{\Phi}_{yy}(z) = \underbrace{\sigma_e \frac{\prod_{i=1}^m (z-r_i)}{\prod_{j=1}^b (z-p_j)}}_{=T(z), \text{ stable, causal}} \cdot \underbrace{\sigma_e \frac{\prod_{i=1}^m (z^{-1}-r_i^*)}{\prod_{j=1}^b (z^{-1}-p_j^*)}}_{=T^*(z^{-*}), \text{ stable, anti-causual}} \;,$$

assuming $|r_i|\langle 1, |p_i|\langle 1, \text{ and } \sigma_e^2 \rangle 0$.



2.2 Additive Decomposition

Let the sequence $\{f_k\}$ have a \mathscr{Z} -transform that exist in an annulus containing the unit circle. Then

$$F(z) = \sum_{k=-\infty}^{\infty} f_k z^{-k} = \underbrace{\sum_{k=0}^{\infty} f_k z^{-k}}_{[F(z)]_+ \text{ casual part }} + \underbrace{\sum_{k=-\infty}^{-1} f_k z^{-k}}_{[E(z)]_- \text{ strictly anti-causal part }}$$

2.3 Solving the Wiener-Hopf equation

Original problem:

$$\xrightarrow{y_k} H^c_{\hat{x}y}(z)$$

 $r_{xy}(\ell+n) = \sum_{i=0}^{\infty} h_i r_{yy}(\ell-i), \quad \forall \ell \ge 0$

New problem:

$$\xrightarrow{y_k} 1/T(z) \xrightarrow{e_k} H^c_{\hat{x}e}(z) \xrightarrow{\hat{x}_{k+n}}$$

Different filter coefficients

$$r_{xe}(\ell+n) = \sum_{i=0}^{\infty} \bar{h}_i r_{ee}(\ell-i) = \sum_{i=0}^{\infty} \bar{h}_i \delta(\ell-i), \quad \forall \ell \ge 0$$

$$\implies \bar{h}_i = \begin{cases} r_{xe}(i+n), & i \ge 0 \\ 0, & i < 0 \end{cases}$$

$$\implies H_{\hat{x}e}^c(z) = \left[\Phi_{xe}(z) z^n \right]_{\perp}$$

Putting it all together

$$w_k = \begin{pmatrix} e_k \\ x_k \end{pmatrix} \longrightarrow \tilde{\Phi}_{ww} = \begin{pmatrix} \tilde{\Phi}_{ee} & \tilde{\Phi}_{ex} \\ \tilde{\Phi}_{xe} & \tilde{\Phi}_{xx} \end{pmatrix}$$
$$u_k = \begin{pmatrix} y_k \\ x_k \end{pmatrix} \longrightarrow \tilde{\Phi}_{uu} = \begin{pmatrix} \tilde{\Phi}_{yy} & \tilde{\Phi}_{yx} \\ \tilde{\Phi}_{xy} & \tilde{\Phi}_{xx} \end{pmatrix}$$

Super formula:

$$\underbrace{u_k} G(z) \xrightarrow{w_k} G(z)$$

$$\check{\Phi}_{ww} = G(z)\check{\Phi}_{ww}G^*(z^{-*})$$

$$G(z) = egin{pmatrix} 1/T(z) & 0 \ 0 & 1 \end{pmatrix} \qquad \Longrightarrow \quad ilde{\Phi}_{xe} = ilde{\Phi}_{xy}/T^*(z^{-*})$$

$$\Longrightarrow H^c_{\hat{x}e}(z) = \left[\frac{z^n \tilde{\Phi}_{xy}(z)}{T^*(z^{-*})}\right]_+ \qquad \Longrightarrow \quad H^c_{\hat{x}y}(z) = \frac{1}{T(z)} \cdot \left[\frac{z^n \tilde{\Phi}_{xy}(z)}{T^*(z^{-*})}\right]_+$$

Note: A factor z^{-n} is in some books added to $H^c_{\hat{x}y}(z)$. Without this factor (as given above) $\hat{x}_{k+n|k} = H^c_{\hat{x}y}(\Delta)y_k$ and with the factor $\hat{x}_{k+n|k} = \Delta^{-n}H^c_{\hat{x}y}(\Delta)y_{k+n}$. Be sure to know which convention you are adhering to, and both works just fine.